

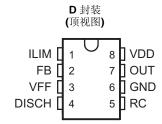
高速电压模式脉宽调制器

查询样品: UCC25705-Q1, UCC25706-Q1

特性

- 符合汽车应用要求
- 工作频率高于 4 MHz
- 集成型振荡器/ 电压前馈 补偿
- 大于 4:1 的输入电压范围
- 25 ns 电流限制延迟
- 可编程最大占空比钳位
- 光耦合器接口
- 50 µA 启动电流
- 1 MHz 时,工作电流为 4.2 mA

- 闭锁电流超过 100mA,符合 JESD78 Class I 标准
- 业界最小占位面积的 8 引脚 MSOP 封装可最大限度地缩减电路板面积与厚度



说明

UCC25705-Q1 与 UCC25706-Q1 器件是具有快速过压保护的 8 引脚电压模式一次侧控制器。上述器件可在高性能隔离与非隔离电源转换器中用作内核高速构建块。

UCC25705-Q1/UCC25706-Q1 器件具有支持集成型前馈补偿的高速振荡器,可提高转换器性能。 针对 25 ns 输出延迟时间的典型电流感测可对过载情况实现快速响应。 此外,该 IC 还可为实现更高保护功能提供可编程最大占空比钳位,其也可针对振荡器进行禁用,在尽可能大的占空比下运行。

提供两个 UVLO 选项。 具有更低启动电压的 UCC25705-Q1 旨在满足 DC 至 DC 转换器的需求,而更高启动电压 及更宽 UVLO 范围的 UCC25706-Q1 则更适用于离线应用。

UCC2570x-Q1 系列采用 8 引脚 SOIC (D) 封装。

图 1. 典型应用原理图 V_{OUT} VDD DISCH FB TPS2829 RC OUT 7 VFF FET DRIVER SOFT GND ILIM 1 CIRCUIT UCC2570x-Q1 MODE =1

A

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ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature (unless otherwise noted)(1)(2)

	VALUE	UNIT
Supply voltage	15	V
Input voltage (VFF,RC,ILIM)	7	V
Input voltage (FB)	15	V
Input current (DISCH)	1	mA
Output current (OUT) dc	±20	mA
Storage temperature, T _{stg}	-65 to 150	°C
Junction temperature, T _J	–55 to 150	°C
Lead temperature (soldering, 10 sec)	300	°C

⁽¹⁾ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		VALUE	UNIT
T _A	Operating ambient temperature	-40 to 105	°C

ORDERING INFORMATION TABLE

T,	A	PAC	KAGE	ORDERABLE PART NUMBER	TOP-SIDE MARKING
40°C to	125°C	SOIC-8 - D	Reel of 2500	UCC25706QDRQ1	25706Q
40°C to	125°C	SOIC-8 - D	Reel of 2500	UCC25705QDRQ1	Preview

ESD RATINGS TABLE

	PARAMETER	VALUE	UNIT
	Human Body Model (HBM)	1000	V
ESD	Charged- Device Model (CDM)	1000	V
	Machine Model (MM)	200	V

⁽²⁾ All voltages are with respect to GND. Currents are positive into, negative out of the specified terminal. Consult ti.com/packaging for more information.



ELECTRICAL CHARACTERISTICS

 V_{DD} = 11 V, V_{IN} = 30 V, R_{T} = 47 k, R_{DISCH} = 400 k, R_{FF} = 14 k, C_{T} = 220 pF, C_{VDD} = 0.1 μ F, and no load on the outputs, T_{A} = -40° to 125°C, (unless otherwise specified)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
UVLO section (UCCx5705)					
Start threshold		8.0	8.8	9.6	V
Stop threshold		7.4	8.2	9.0	V
Hysteresis		0.3	0.6	1.0	V
UVLO section (UCCx5706)				•	
Start threshold		11.2	12.0	12.8	V
Stop threshold		7.2	8.0	8.8	V
Hysteresis		3.2	4.0	4.5	V
Supply Current Section					
Start-up current	V _{DD} = UVLO start – 1 V, V _{DD} comparator off		30	90	μΑ
I _{DD} active	V _{DD} comparator on, oscillator running at 1 MHz		4.2	5.0	mA
Line Sense Section					
Low line comparator threshold		0.95	1.00	1.15	V
Input bias current (VFF)		-100		100	nA
Oscillator Section		*			
Frequency	VFF = 1.2 V to 4.8 V	0.9	1.0	1.1	MHz
CT peak voltage	VFF = 1.2 V, See ⁽¹⁾		1.2		V
	VFF = 4.8 V, See ⁽¹⁾		4.8		V
CT valley voltage	See ⁽¹⁾		0		V
Current Limit Section				•	
Input bias current		0.2	-0.2	-1	μΑ
Current limit threshold		180	200	220	mV
Propagation delay, ILIM to OUT	50 mV overdrive		25	35	ns
Pulse Width Modulator Section					
FB input impedance	V _{FB} = 3 V	30	50	90	kΩ
Minimum duty cycle	V _{FB} < 2 V			0	%
Navigas yan districtive and a	$V_{FB} = V_{DD}$, $F_{OSC} = 1 \text{ MHz}$	70	75	80	%
Maximum duty cycle	$V_{DISCH} = 0 V$, $F_{OSC} = 1 MHz$		93		%
PWM gain	V _{FF} = 2.5 V, MODE = 1		12		%/V
Propagation delay, PWM to OUT			65	130	ns
Output Section					
V _{OH}	$I_{OUT} = -5 \text{ mA}, \qquad V_{DD} - \text{output}$		0.3	0.6	V
V _{OL}	I _{OUT} = 5 mA		0.15	0.4	V
Rise time	C _{LOAD} = 50 pF		10	25	ns
Fall time	C _{LOAD} = 50 pF		10	25	ns

⁽¹⁾ Specified by design.



PIN DESCRIPTIONS

DISCH: A resistor to VIN sets the oscillator discharge current programming a maximum duty cycle. When grounded, an internal comparator switches the oscillator to a quick discharge mode. A small 100-pF capacitor between DISCH and GND may reduce oscillator jitter without impacting feed-forward performance. I_{DISCH} must be between 25 μ A and 250 μ A over the entire V_{IN} range.

FB: Input to the PWM comparator. This pin is intended to interface with an optocoupler. Input impedance is $50-k\Omega$ typical.

GND: Ground return pin.

I_{LIM}: Provides a pulse-by-pulse current limit by terminating the PWM pulse when the input is above 200 mV. This provides a high speed (25 ns typical) path to reset the PWM latch, allowing for a pulse-by-pulse current limit.

OUT: The output is intended to drive an external FET driver or other high impedance circuits, but is not intended to directly drive a power MOSFET. This improves the controller's noise immunity. The output resistance of the PWM controller, typically $60~\Omega$ pull-up and $30~\Omega$ pull-down, will result in excessive rise and fall times if a power MOSFET is directly driven at the speeds for which the UCC2570x-Q1 is optimized.

RC: The oscillator can be configured to provide a maximum duty cycle clamp. In this mode the on-time is set by RT and CT, while the off-time is set by R_{DISCH} and CT.Since the voltage ramp on CTis proportional to VIN, feed-forward action is obtained. Since the peak oscillator voltage is also proportional to VIN, constant frequency operation is maintained over the full power supply input range. When the DISCH pin is grounded, the duty cycle clamp is disabled. The RC pin then provides a low impedance path to ground CT during the off time.

 V_{DD} : Power supply pin. This pin should be bypassed with a 0.1- μ F capacitor for proper operation. The undervoltage lockout function of the UCC2570x-Q1 allows for a low current startupmode and ensures that all circuits become active in a known state. The UVLO thresholds on the UCC25705-Q1 are appropriate for a dc-to-dc converter application. The wider UVLO hysteresis of the UCC25706-Q1 (typically 4 V) is optimized for a bootstrap startup mode from a high impedance source.

 V_{FF} : The feed-forward pin provides the controllerwith a voltage proportional to the power supply input voltage. When the oscillator is providing a duty cycle clamp, a current of 2 × I_{DISCH} is sourced from the V_{FF} pin. A single resistor R_{FF} between V_{FF} and GND then set V_{FF} to:

$$VFF \approx VIN \times \left(\frac{2 \times R_{FF}}{2 \times R_{FF} + R_{DISCH}} \right)$$

When the DISCH pin is grounded and the duty cycle clamp is not used, the internal current source is disabled and a resistor divider from VIN is used to set VFF. In either case, when the voltage on V_{FF} is less than 1.0 V, both the output and oscillator are disabled.

STRUMENTS

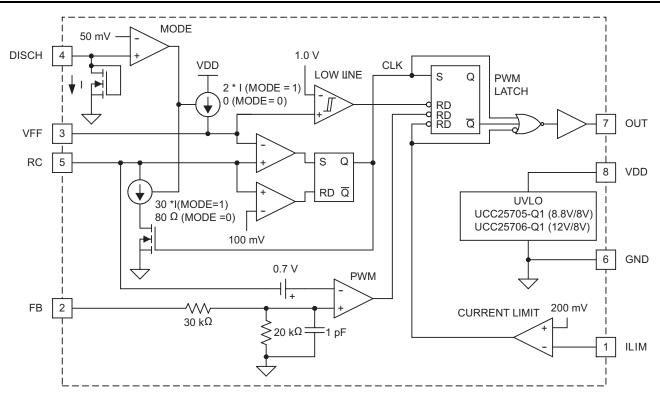


Figure 2. Block Diagram

FUNCTIONAL DESCRIPTION

Oscillator and PWM

The oscillator can be programmed to provide a duty cycle clamp or be configured to run at the maximum possible duty cycle.

The PWM latch is set during the oscillator discharge and is reset by the PWM comparator when the C_T waveform is greater than the feedback voltage. The voltage at the FB pin is attenuated before it is applied to the PWM comparator. The oscillator ramp is shifted by approximately 0.65-V at room temperature at the PWM comparator. The offset has a temperature coefficient of approximately --2 mV/ $^{\circ}$ C.

The I_{LIM} comparator adds a pulse by pulse current limit by resetting the PWM latch when $V_{ILIM} > 200$ mV. The PWM latch is also reset by a low line condition ($V_{FF} < 1.0$ V).

All reset conditions are dominant; asserting any output will force a zero duty cycle output.

Oscillator With Duty Cycle Clamp (MODE = 1)

The timing capacitor C_T is charged from ground to V_{FF} through R_T . The discharge path is through an on-chip current sink that has a value of 30 × I_{DISCH} , where I_{DISCH} is the current through the external resistor R_{DISCH} . Since the charge and discharge currents are both proportional to V_{IN} , their ratio, and the maximum duty cycle remains constant as V_{IN} varies.





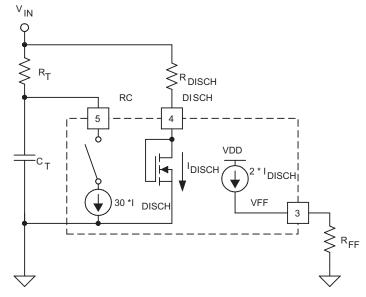


Figure 3. Duty Cycle Clamp (MODE = 1)

The on-time is approximately:

$$T_{\text{ON}} = \alpha \times R_{\text{T}} \times C_{\text{T}} \text{ where } \alpha = \frac{V_{\text{FF}}}{V_{\text{IN}}} \approx \frac{2 \times R_{\text{FF}}}{R_{\text{DISCH}}}$$

The off-time is:

$$T_{\text{OFF}} = \alpha \times \frac{C_{\text{T}} \times \left(R_{\text{T}} \times R_{\text{DISCH}}\right)}{30 \times R_{\text{T}} - R_{\text{DISCH}}}$$

The frequency is:

$$f = \frac{1}{\alpha \times R_T \times C_T} \times \frac{1}{1 + \frac{R_{DISCH}}{30 \times R_T - R_{DISCH}}}$$

The maximum duty cycle is:

$$Duty\,Cycle = \frac{T_{ON}}{T_{ON} + T_{OFF}} = \left(1 - \frac{R_{DISCH}}{30 \times R_T}\right)$$



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Component Selection for Oscillator With Duty Cycle Clamp (MODE = 1)

For a power converter with the following specifications:

- $V_{IN(min)} = 18 \text{ V}$
- $V_{IN(max)} = 75 \text{ V}$
- $V_{IN(shutdown)} = 15 \text{ V}$
- $F_{OSC} = 1 MHz$
- MAX = 0.78 at $V_{IN(min)}$

In this mode, the on-time is approximately:

- $T_{ON(max)} = 780 \text{ ns}$
- $T_{OFF(min)} = 220 \text{ ns}$

$$V_{FF(min)} = \frac{18}{15} = 1.20 \text{ V}$$

- 1. Pick $C_T = 220 pF$.
- 2. Calculate R_T.

$$R_{\text{T}} = \frac{V_{\text{IN(min)}} \! \times \! T_{\text{ON(max)}}}{V_{\text{FF(min)}} \! \times \! C_{\text{T}}}$$

$$R_T = 51.1 \text{ k}\Omega$$

 R_{DISCH}

$$R_{DISCH} = \frac{30 \times R_{T}}{1 + \left(\frac{\left(\frac{V_{FF(min)}}{V_{IN(min)}}\right) \times R_{T} \times C_{T}}{T_{OFF(min)}}\right)}$$

 $R_{DISCH} = 383 \text{ k}\Omega.$

 I_{DISCH} must be between 25 μA and 250 μA over the entire VIN range.

With the calculated values, I_{DISCH} ranges from 44 µA to 193 µA, within the allowable range. If I_{DISCH} is too high, C_T must be decreased.

4.
$$R_{FF}$$

$$R_{FF} = \frac{V_{FF(min)} \times R_{DISCH}}{2 \times (V_{IN(min)} - 1)}$$

The nearest 1% standard value to the calculated value is 13.7 k.

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Oscillator Without Duty Cyle Clamp (MODE = 0)

In this mode, the timing capacitor is discharged through a low impedance directly to ground. The DISCH pin is externally grounded. A comparator connected to DISCH senses the ground connection and disables both the discharge current source and V_{FF} current source. A resistor divider is now required to set V_{FF}.

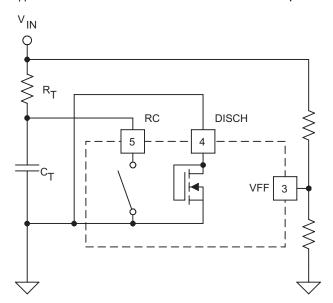


Figure 4. Ocsillator Without Clamp (MODE = 0)

$$T_{\text{ON}} = \alpha \times R_{\text{T}} \times C_{\text{T}} \text{ where } \alpha = \frac{V_{\text{FF}}}{V_{\text{IN}}}$$
 In this mode, the on-time is approximately:

The off-time is: $T_{OFF} \approx 75 \text{ ns}$

The frequency is:

$$f = \frac{1}{\alpha \! \times \! R_{\scriptscriptstyle T} \! \times \! C_{\scriptscriptstyle T} \! + \! 75 ns}$$

NSTRUMENTS



Component Selection for Oscillator Without Duty Cycle Clamp (MODE = 0)

For a power converter with the following specifications:

- V_{IN(min)} = 18 V
- V_{IN(max)} = 75 V
- V_{IN(shutdown)} = 15 V
- F_{OSC} = 1 MHz

With these specifications,

$$V_{\text{FF(min)}} = \frac{18}{15} = 1.2V$$

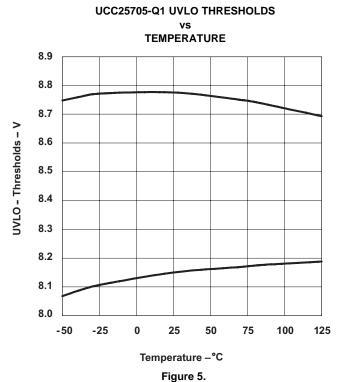
- 1. Pick $C_T = 220 pF$
- 2. Calculate R_T.

$$R_{T} = \frac{\frac{V_{\text{IN(min)}}}{V_{\text{FF(min)}}} \times \left(\frac{1}{F_{\text{OSC}}} - 75 ns\right)}{C_{T}}$$

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TEXAS INSTRUMENTS

TYPICAL CHARACTERISTICS



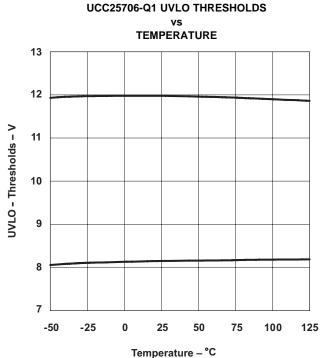
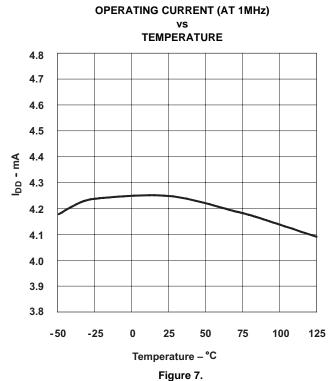


Figure 6.



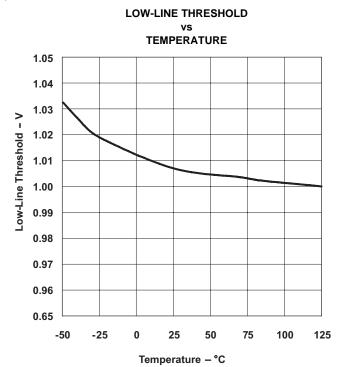
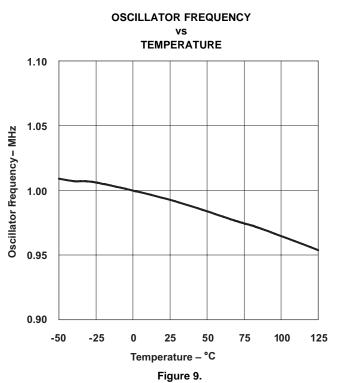
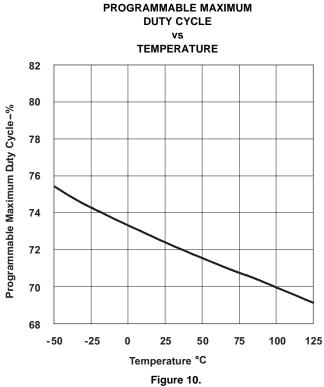
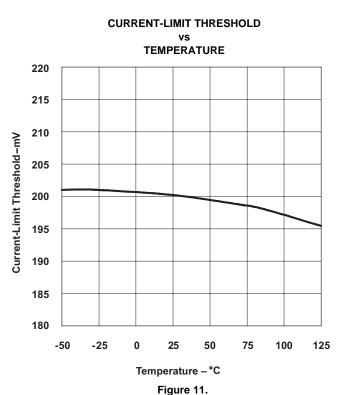


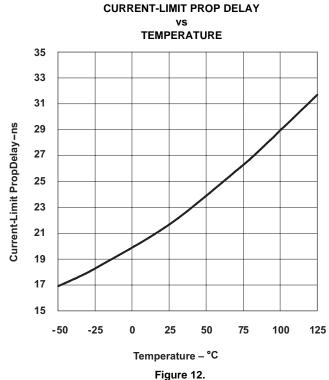
Figure 8.

TYPICAL CHARACTERISTICS









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放大器和线性器件	http://www.ti.com.cn/amplifiers	计算机及周边	www.ti.com.cn/computer
数据转换器	http://www.ti.com.cn/dataconverters	消费电子	www.ti.com/consumer-apps
DLP®产品	www.dlp.com	能源	www.ti.com/energy
DSP - 数字信号处理器	http://www.ti.com.cn/dsp	工业应用	www.ti.com.cn/industrial
时钟和计时器	http://www.ti.com.cn/clockandtimers	医疗电子	www.ti.com.cn/medical
接口	http://www.ti.com.cn/interface	安防应用	www.ti.com.cn/security
逻辑	http://www.ti.com.cn/logic	汽车电子	www.ti.com.cn/automotive
电源管理	http:///www.ti.com.cn/power	视频和影像	www.ti.com.cn/video
微控制器 (MCU)	http://www.ti.com.cn/microcontrollers	无线通信	www.ti.com.cn/wireless
RFID 系统	http://www.ti.com.cn/rfidsys		
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PACKAGE OPTION ADDENDUM

10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
UCC25706QDRQ1	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	25706Q	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

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- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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