TLV5628C, TLV5628I OCTAL 8-BIT DIGITAL-TO-ANALOG CONVERTERS

DACB [

DACA [

DATA [

CLK [

DACE 7

DACF [] 8

GND ∏ 3

4

V_{DD} [] 6

SLAS108A - JANUARY 1995 - REVISED NOVEMBER 1995

16 DACC 15 DACD

14 REF1

13 LDAC

12 LOAD

11 | REF2

10 DACH

9 DACG

DW OR N PACKAGE (TOP VIEW)

- Eight 8-Bit Voltage Output DACs
- 3-V Single Supply Operation
- Serial Interface
- High-Impedance Reference Inputs
- Programmable for 1 or 2 Times Output Range
- Simultaneous Update Facility
- Internal Power-On Reset
- Low Power Consumption
- Half-Buffered Output

applications

- Programmable Voltage Sources
- Digitally Controlled Amplifiers/Attenuators
- Mobile Communications
- Automatic Test Equipment
- Process Monitoring and Control
- Signal Synthesis

description

The TLV5628C and TLV5628I are octal 8-bit voltage output digital-to-analog converters (DACs) with buffered reference inputs (high impedance). The DACs produce an output voltage that varies between one or two times the reference voltages and GND, and the DACs are monotonic. The device is simple to use, running from a single supply of 3 to 3.6 V. A power-on reset function is incorporated to ensure repeatable start-up conditions.

Digital control of the TLV5628C and TLV5628I is over a simple 3-wire serial bus that is CMOS compatible and easily interfaced to all popular microprocessor and microcontroller devices. The 12-bit command word comprises 8 bits of data, 3 DAC select bits and a range bit, the latter allowing selection between the times 1 or times 2 output range. The DAC registers are double buffered, allowing a complete set of new values to be written to the device, then all DAC outputs are updated simultaneously through control of the LDAC terminal. The digital inputs feature Schmitt triggers for high noise immunity.

The 16-terminal small-outline D package allows digital control of analog functions in space-critical applications. The TLV5628C is characterized for operation from 0°C to 70°C. The TLV5628I is characterized for operation from –40°C to 85°C. The TLV5628C and TLV5628I do not require external trimming.

AVAILABLE OPTIONS

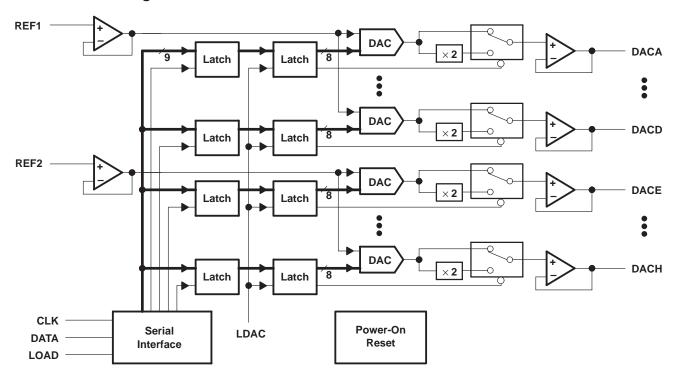
PACKAGE								
TA	SMALL OUTLINE (DW)	PLASTIC DIP (N)						
0°C to 70°C	TLV5628CDW	TLV5628CN						
-40°C to 85°C	TLV5628IDW	TLV5628IN						



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



functional block diagram



Terminal Functions

TERMIN	IAL	1/0	DESCRIPTION					
NAME	NO.	1/0	DESCRIPTION					
CLK	5	Ι	Serial-interface clock, data enters on the negative edge					
DACA	2	0	DACA analog output					
DACB	1	0	DACB analog output					
DACC	16	0	DACC analog output					
DACD	15	0	DACD analog output					
DACE	7	0	DACE analog output					
DACF	8	0	DACF analog output					
DACG	9	0	DACG analog output					
DACH	10	0	DACH analog output					
DATA	4	Ι	Serial-interface digital data input					
GND	3	I	Ground return and reference terminal					
LDAC	13	Ι	DAC-update latch control					
LOAD	12	Ι	Serial-interface load control					
REF1	14	I	Reference voltage input to DACA					
REF2	11	Ι	Reference voltage input to DACB					
V_{DD}	6	Ι	Positive supply voltage					



detailed description

The TLV5628 is implemented using eight resistor-string DACs. The core of each DAC is a single resistor with 256 taps, corresponding to the 256 possible codes listed in Table 1. One end of each resistor string is connected to the GND terminal and the other end is fed from the output of the reference input buffer. Monotonicity is maintained by use of the resistor strings. Linearity depends upon the matching of the resistor elements and upon the performance of the output buffer. Because the inputs are buffered, the DACs always present a high-impedance load to the reference sources. There are two input reference terminals; REF1 is used for DACA through DACD and REF2 is used by DACE through DACH.

Each DAC output is buffered by a configurable-gain output amplifier, which can be programmed to times 1 or times 2 gain.

On power-up, the DACs are reset to CODE 0.

Each output voltage is given by:

$$V_O(DACA|B|C|D|E|F|G|H) = REF \times \frac{CODE}{256} \times (1 + RNG bit value)$$

where CODE is in the range of 0 to 255 and the range (RNG) bit is a 0 or 1 within the serial-control word.

data interface

With LOAD high, data is clocked into the DATA terminal on each falling edge of CLK. Once all data bits have been clocked in, LOAD is pulsed low to transfer the data from the serial-input register to the selected DAC as shown in Figure 1. When LDAC is low, the selected DAC output voltage is updated and LOAD goes low. When LDAC is high during serial programming, the new value is stored within the device and can be transferred to the DAC output at a later time by pulsing LDAC low as shown in Figure 2. Data is entered MSB first. Data transfers using two 8 clock cycle periods are shown in Figures 3 and 4.

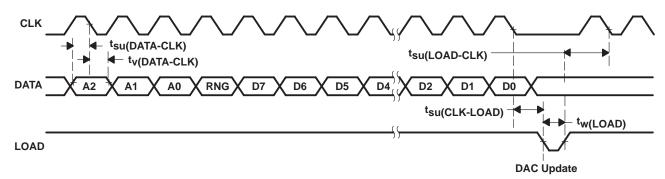


Figure 1. LOAD-Controlled Update (LDAC = Low)

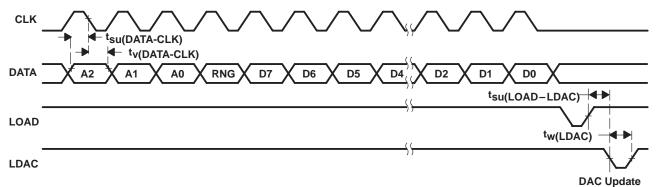


Figure 2. LDAC-Controlled Update



LOAD

LDAC

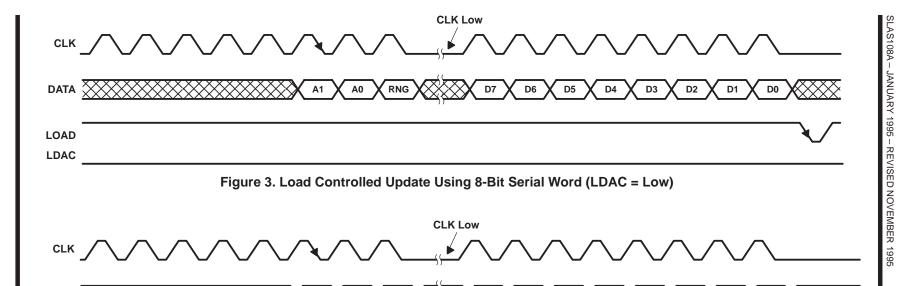


Figure 4. LDAC Controlled Update Using 8-Bit Serial Word

data interface (continued)

Table 2 lists the A2, A1, and A0 bits and the selection of the updated DACs. The RNG bit controls the DAC output range. When RNG = low, the output range is between the applied reference voltage and GND, and when RNG = high, the range is between twice the applied reference voltage and GND.

Table 1. Ideal Output Transfer

D7	D6	D5	D4	D3	D2	D1	D0	OUTPUT VOLTAGE
0	0	0	0	0	0	0	0	GND
0	0	0	0	0	0	0	1	(1/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
0	1	1	1	1	1	1	1	(127/256) × REF (1+RNG)
1	0	0	0	0	0	0	0	(128/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
1	1	1	1	1	1	1	1	(255/256) × REF (1+RNG)

Table 2. Serial Input Decode

A2	A1	A0	DAC UPDATED
0	0	0	DACA
0	0	1	DACB
0	1	0	DACC
0	1	1	DACD
1	0	0	DACE
1	0	1	DACF
1	1	0	DACG
1	1	1	DACH

linearity, offset, and gain error

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset the output voltage may not change with the first code depending on the magnitude of the offset voltage.

The output amplifier, with a negative voltage offset, attempts to drive the output to a negative voltage. However, since the most negative supply rail is ground, the output cannot drive to a negative voltage.

So when the output offset voltage is negative, the output voltage remains at 0 volts until the input code value produces a sufficient output voltage to overcome the inherent negative offset voltage resulting in the transfer function shown in Figure 5.

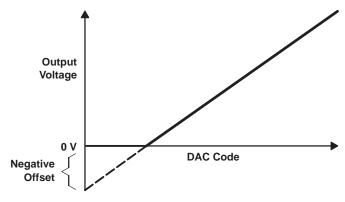


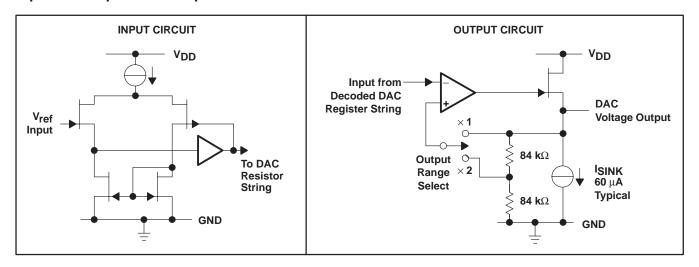
Figure 5. Effect of Negative Offset (Single Supply)

The negative offset error produces a breakpoint, not a linearity error. The transfer function would follow the dotted line if the output buffer could drive to a negative voltage.

For a DAC, linearity is measured between zero input code (all inputs 0) and full scale code (all inputs 1) after offset and full scale is adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. The linearity in the unipolar mode is measured between full scale code and the lowest code which produces a positive output voltage.

The code is calculated from the maximum specification for the negative offset.

equivalent inputs and outputs





absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage (V _{DD} – GND)	
Digital input voltage range, V _{ID}	
Reference input voltage range	GND $- 0.3 \text{ V to V}_{DD} + 0.3 \text{ V}$
Operating free-air temperature range, T _A : TLV5628C	0°C to 70°C
TLV5628I	40°C to 85°C
Storage temperature range, T _{stg}	–50°C to 150°C
Storage temperature range, T_{stg}	230°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V _{DD}		2.7	3.3	5.25	V
High-level digital input voltage, VIH		0.8 V _{DD}			V
Low-level digital input voltage, V _{IL}				0.8	V
Reference voltage, V _{ref} [A B C D E F G H], X1	gain			V _{DD} -1.5	V
Load resistance, R _L		10			kΩ
Setup time, data input, t _{SU(DATA-CLK)} (see Fi	gures 1 and 2)	50			ns
Valid time, data input valid after CLK↓, t _{V(DATA-CLK)} (see Figures 1 and 2) 50					ns
Setup time, CLK eleventh falling edge to LOAI	50			ns	
Setup time, LOAD↑ to CLK↓, t _{SU} (LOAD-CLK)	(see Figure 1)	50			ns
Pulse duration, LOAD, t _{W(LOAD)} (see Figure	1)	250			ns
Pulse duration, LDAC, tw(LDAC) (see Figure 2	2)	250			ns
Setup time, LOAD↑ to LDAC↓, t _{SU(LOAD-LDA}	(C) (see Figure 2)	0			ns
CLK frequency				1	MHz
Operating free-air temperature, T _A	TLV5628C	0		70	°C
Operating free-all temperature, 1A	TLV5628I	-40		85	°C

electrical characteristics over recommended operating free-air temperature range, V_{DD} = 3 V to 3.6 V, V_{ref} = 2 V, \times 1 gain output range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
lн	High-level digital input current	$V_I = V_{DD}$			±10	μΑ
I _I L	Low-level digital input current	V _I = 0 V			±10	μΑ
IO(sink)	Output sink current	Foob DAC output	20			μΑ
IO(source)	Output source current	Each DAC output	1			mA
C.	Input capacitance			15		n.E
Ci	Reference input capacitance			15		pF
I _{DD}	Supply current	V _{DD} = 3.3 V			4	mA
I _{ref}	Reference input current	$V_{DD} = 3.3 \text{ V}, V_{ref} = 1.5 \text{ V}$			±10	μΑ
EL	Linearity error (end point corrected)	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 1)}$			±1	LSB
ED	Differential linearity error	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 2)}$			±0.9	LSB
EZS	Zero-scale error	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 3)}$	0		30	mV
	Zero-scale error temperature coefficient	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 4)}$		10		μV/°C
E _{FS}	Full-scale error	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 5)}$			±60	mV
	Full-scale error temperature coefficient	$V_{ref} = 1.25 \text{ V}, \times 2 \text{ gain (see Note 6)}$		±25		μV/°C
PSRR	Power supply sensitivity	See Notes 7 and 8		0.5		mV/V

- NOTES: 1. Integral nonlinearity (INL) is the maximum deviation of the output from the line between zero-scale and full scale (excluding the effects of zero code and full-scale errors).
 - 2. Differential nonlinearity (DNL) is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.
 - 3. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.
 - 4. Zero-scale error temperature coefficient is given by: $ZSETC = [ZSE(T_{max}) ZSE(T_{min})]/V_{ref} \times 10^6/(T_{max} T_{min})$.
 - 5. Full-scale error is the deviation from the ideal full-scale output ($V_{ref} 1$ LSB) with an output load of 10 k Ω .
 - 6. Full-scale temperature coefficient is given by: FSETC = [FSE(T_{max}) FSE (T_{min})]/V_{ref} × 10⁶/(T_{max} T_{min}).
 - Zero-scale error rejection ratio (ZSE-RR) is measured by varying the V_{DD} voltage from 4.5 V to 5.5 V dc and measuring the effect
 of this signal on the zero-code output voltage.
 - Full-scale error rejection ratio (FSE-RR) is measured by varing the V_{DD} voltage from 3 V to 3.6 V dc and measuring the effect of this signal on the full-scale output voltage.

operating characteristics over recommended operating free-air temperature range, V_{DD} = 3 V to 3.6 V, V_{ref} = 2 V, \times 1 gain output range (unless otherwise noted)

	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output slew rate	$C_L = 100 \text{ pF}, R_L = 10 \text{ k}\Omega$		1		V/µs
Output settling time	To 0.5 LSB, $C_L = 100 \text{ pF}$, $R_L = 10 \text{ k}\Omega$, See Note 9		10		μs
Large-signal bandwidth	Measured at -3 dB point		100		kHz
Digital crosstalk	CLK = 1-MHz square wave measured at DACA-DACH		-50		dB
Reference feedthrough	See Note 10		-60		dB
Channel-to-channel isolation	See Note 11		-60		dB
Reference input bandwidth	See Note 12		100		kHz

- NOTES: 9. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 00 hex to FF hex or FF hex to 00 hex. For TLC5628C V_{DD} = 5 V, V_{ref} = 2 V and range = ×2. For TLC5628I V_{DD} = 3 V, V_{ref} = 1.25 V and range ×2.
 - 10. Reference feedthrough is measured at any DAC output with an input code = 00 hex with a V_{ref} input = 1 V dc + 1 Vpp at 10 kHz.
 - 11. Channel-to-channel isolation is measured by setting the input code of one DAC to FF hex and the code of all other DACs to 00 hex with V_{ref} input = 1 V dc + 1 Vpp at 10 kHz.
 - 12. Reference bandwidth is a -3 dB bandwidth with an input at V_{ref} = 1.25 V dc + 2 V_{PP} and with a full-scale digital input code.



PARAMETER MEASUREMENT INFORMATION

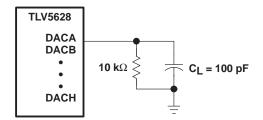


Figure 6. Slewing Settling Time and Linearity Measurements

TYPICAL CHARACTERISTICS

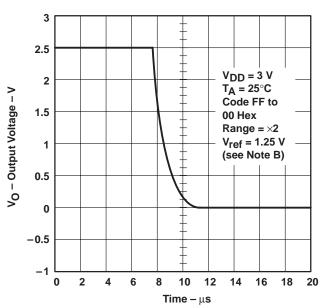
POSITIVE RISE TIME AND SETTLING TIME

3 2.5 2 V_O - Output Voltage - V 1.5 1 $V_{DD} = 3 V$ $T_A = 25^{\circ}C$ 0.5 Code 00 to FF Hex Range = \times 2 0 V_{ref} = 1.25 V (see Note A) -0.5-1 0 2 4 6 8 10 12 14 16 18 20 Time – µs

NOTE A: Rise time = 2.05 μs , positive slew rate = 0.96 V/ μs , settling time = 4.5 μs .

Figure 7

NEGATIVE FALL TIME AND SETTLING TIME

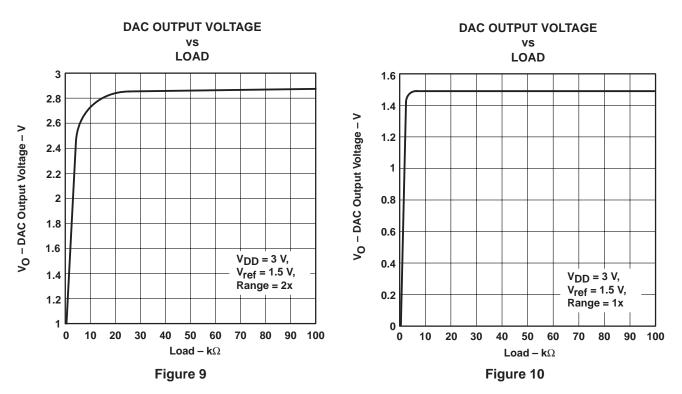


NOTE B: Fall time = 4.25 μs , negative slew rate = 0.46 V/ μs , settling time = 8.5 μs .

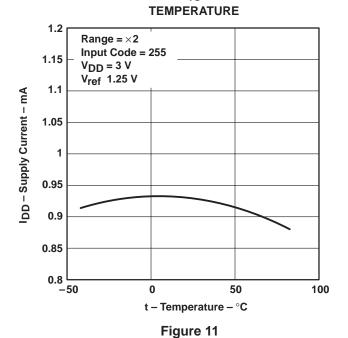
Figure 8



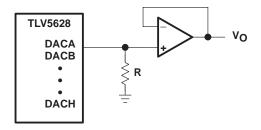
TYPICAL CHARACTERISTICS



SUPPLY CURRENT vs



APPLICATION INFORMATION



NOTE A: Resistor R \geq 10 k Ω

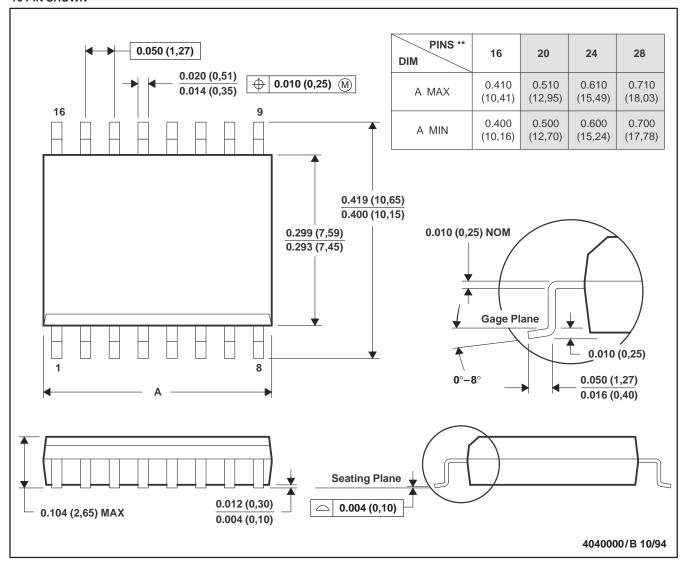
Figure 12. Output Buffering Scheme

MECHANICAL DATA

DW (R-PDSO-G**)

16 PIN SHOWN

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES: A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
 - D. Falls within JEDEC MS-013

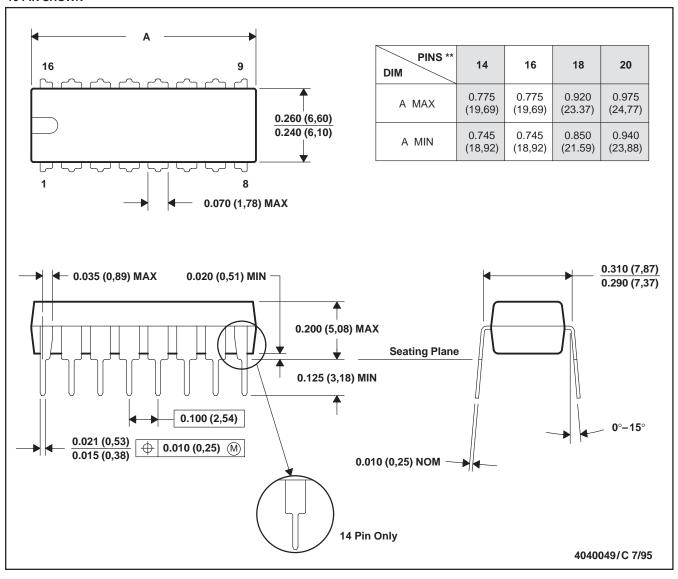


MECHANICAL DATA

N (R-PDIP-T**)

16 PIN SHOWN

PLASTIC DUAL-IN-LINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-001 (20-pin package is shorter than MS-001)





10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing		Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
TLV5628CDW	ACTIVE	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLV5628C	Samples
TLV5628CDWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM		TLV5628C	Samples
TLV5628CN	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type		TLV5628CN	Samples
TLV5628IDW	ACTIVE	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLV5628I	Samples
TLV5628IDWG4	ACTIVE	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLV5628I	Samples
TLV5628IDWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM		TLV5628I	Samples
TLV5628IDWRG4	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM		TLV5628I	Samples
TLV5628IN	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type		TLV5628IN	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

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⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



PACKAGE OPTION ADDENDUM

10-Dec-2020

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5628CDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV5628CDWR	SOIC	DW	16	2000	350.0	350.0	43.0

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