











INA290-Q1

ZHCSKF9-OCTOBER 2019

符合 AEC-Q100 标准的 INA290-Q1 2.7V 至 110V、800kHz、超精密小型 (SC-70) 封装电流感应放大器

1 特性

- 符合面向汽车 应用的 AEC-Q100 标准:
 - 温度等级 1: -40°C 至 +125°C, T_A
- 宽共模电压:
 - 工作电压: 2.7V 至 110V
 - 可承受电压: -20V 至 +120V
- 出色的共模抑制比 (CMRR):
 - 160dB 直流
 - 85dB 交流 (50kHz 时)
- 精度
 - 增益:
 - 增益误差: 0.25% (最大值)
 - 增益漂移: 10ppm/°C(最大值)
 - 失调电压:
 - 失调电压: ±25µV (最大值)
 - 温漂: ±0.25µV/°C(最大值)
- 可用增益:
 - INA290A1Q: 20V/V
 - INA290A2Q: 50V/V
 - INA290A3Q: 100V/V
 - INA290A4Q: 200V/V
 - INA290A5Q: 500V/V
- 帯宽: 800kHz
- 压摆率: 2V/µs
- 静态电流: 350µA

2 应用

- 有源天线系统 mMIMO (AAS)
- 宏远程无线电单元 (RRU)
- 48V 机架式服务器
- 48V 商用网络和服务器电源 (PSU)
- 48V 电池管理系统 (BMS)

3 说明

INA290-Q1 是一款超精密电流检测放大器,,在 2.7V 至 110V 的宽共模范围内测量分流电阻器上的压降。它 采用节省空间的 SC-70 封装, PCB 的空间占用仅为 2.00mm × 1.25mm。该器件在 25μV(最大值)的超低失调电压、0.25%(最大值)的小增益误差和 160dB(典型值)的高直流 CMRR 等特性的综合作用下,可实现超高精度的电流测量精度。INA290-Q1 不仅适用于直流电流测量,还适用于高速 应用(例如,快速过流保护),其具有 800kHz(增益为 20V/V)的高带宽和 85dB(50kHz 时)的交流 CMRR。

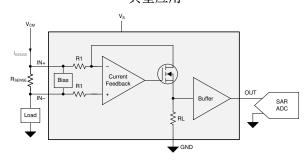
INA290-Q1 由 2.7V 至 20V 单电源供电,消耗的电源电流为 350 μ A(典型值)。INA290-Q1 提供五种增益选项: 20V/V、50V/V、100V/V、200V/V 和 500V/V。零漂移架构的低失调电压能够在扩展工作温度范围(-40° C 至 $+125^{\circ}$ C)内使用低欧姆值分流器进行电流检测。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
INA290-Q1	SC-70 (5)	2.00mm × 1.25mm

(1) 如需了解所有可用封装,请参阅数据表末尾的封装选项附录。

典型应用





目录

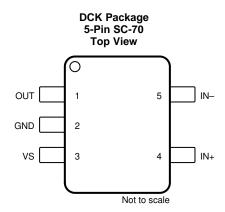
特性 1	8 Application and Implementation	9
应用 1	8.1 Application Information	9
	8.2 Typical Application	10
	9 Power Supply Recommendations	11
	10 Layout	11
	10.1 Layout Guidelines	11
•	10.2 Layout Example	11
3	11 器件和文档支持	12
6.3 Recommended Operating Conditions 4	11.1 文档支持	12
6.4 Thermal Information	11.2 接收文档更新通知	12
6.5 Electrical Characteristics 4		
7.1 Overview 6	11.5 静电放电警告	12
7.2 Functional Block Diagram 6	11.6 Glossary	12
7.3 Feature Description 7	12 机械、封装和可订购信息	12
7.4 Device Functional Modes 8		
	说明 1 修订历史记录 2 Pin Configuration and Functions 3 Specifications 3 6.1 Absolute Maximum Ratings 3 6.2 ESD Ratings 3 6.3 Recommended Operating Conditions 4 6.4 Thermal Information 4 6.5 Electrical Characteristics 4 Detailed Description 6 7.1 Overview 6 7.2 Functional Block Diagram 6 7.3 Feature Description 7	应用 1 8.1 Application Information 说明 1 8.2 Typical Application 修订历史记录 2 9 Power Supply Recommendations Pin Configuration and Functions 3 10 Layout Specifications 3 10.1 Layout Guidelines 6.1 Absolute Maximum Ratings 3 10.2 Layout Example 6.2 ESD Ratings 3 11 器件和文档支持 6.3 Recommended Operating Conditions 4 11.1 文档支持 6.4 Thermal Information 4 11.2 接收文档更新通知 6.5 Electrical Characteristics 4 11.3 支持资源 Detailed Description 6 11.4 商标 7.1 Overview 6 11.5 静电放电警告 7.2 Functional Block Diagram 6 11.6 Glossary 7.3 Feature Description 7 12 机械、封装和可订购信息

4 修订历史记录

日期	修订版本	说明
2019年10月	*	初始发行版



5 Pin Configuration and Functions



Pin Functions

PIN		TVDE	DESCRIPTION	
NAME	NO.	TYPE	DESCRIPTION	
GND	2	Ground	Ground	
OUT	1	Output Output voltage		
VS	3	Power	Power supply	
IN+	4	Input	Connect to supply side of shunt resistor	
IN-	5	Input	Connect to load side of shunt resistor	

Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
Supply Voltage (V _s)			22	V
Analog Inputs,	Differential (V _{IN+}) - (V _{IN-})	-12	12	V
Analog Inputs, V _{IN+} , V _{IN-} (2)	Common - mode	-20	120	V
Output		GND - 0.3	Vs + 0.3	V
T _A	Operating Temperature	- 55	150	°C
TJ	Junction temperature		150	°C
T _{stg}	Storage temperature	-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

	-		VALUE	UNIT
V	Electrostatio discherre	Human body model (HBM), per AEC Q100-002, all pins ⁽¹⁾ HBM ESD Classification Level 2	±2000	V
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per AEC Q100- 011, all pins CDM ESD Classification Level C6	±1000	V

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

VIN+ and VIN- are the voltages at the VIN+ and VIN- pins, respectively.



6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{CM}	Common-mode input range	Vs	48	110	V
Vs	Operating supply range	2.7	5	20	V
T _A	Ambient temperature	-40		125	°C

6.4 Thermal Information

		INA290-Q1	
	THERMAL METRIC ⁽¹⁾	DCK (SC-70)	UNIT
		PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	191.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	144.4	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	69.2	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	46.2	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	69.0	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.5 Electrical Characteristics

at $T_A = 25$ °C, $V_S = 5$ V, $V_{SENSE} = V_{IN+} - V_{IN-} = 0.5$ V / Gain, $V_{CM} = V_{IN-} = 48$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT						
CMRR	Common-mode rejection ratio	V _S < V _{CM} < 110 V, T _A = -40°C to +125°C	140	160		dB
		f = 50 kHz		85		dB
V _{os}	Offset voltage, input referred			5	25	μV
dV _{os} /dT	Offset voltage drift	$T_A = -40$ °C to 125°C			0.25	μV/°C
PSRR	Power supply rejection ratio, input reffered	V _S = 2.7 V to 20 V, T _A = -40°C to +125°C		0.1	1	μV/V
I _B	Input bias current	I _{B+} , V _{SENSE} = 0 mV		20		uA
I _B	Input bias current	I _{B-} , V _{SENSE} = 0 mV		20		uA
OUTPUT						
		INA290A1Q		20		V/V
		INA290A2Q		50		V/V
G	Gain	INA290A3Q		100		V/V
		INA290A4Q		200		V/V
		INA290A5Q		500		V/V
(Caia array	$GND + 50 \; mV \leq V_{OUT} \leq V_{S} - 200 \; mV$		0.02	0.25	%
G _{ERR}	Gain error	$T_A = -40$ °C to $+125$ °C		1	10	ppm/°C
	Nonlinearity error			0.01		%
	Maximum capacitive load	No sustained oscillations, no isolation resistor		1		nF



Electrical Characteristics (continued)

at $T_A = 25$ °C, $V_S = 5$ V, $V_{SENSE} = V_{IN+} - V_{IN-} = 0.5$ V / Gain, $V_{CM} = V_{IN-} = 48$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT		
VOLTAG	E OUTPUT						
	Swing to Vs Power Supply Rail	R_{LOAD} = 10 k Ω to GND, T_A = -40°C to +125°C	Vs - 0.07	Vs - 0.2	V		
	Swing to Ground	R_{LOAD} = 10 k Ω to GND, V_{SENSE} = 0 mV, T_A = -40°C to +125°C	VGND + 0.005	VGND + 0.025	V		
FREQUE	NCY RESPONSE			,			
BW	Bandwidth	INA290A1Q, C _{LOAD} = 5 pF, V _{SENSE} = 200mV	800		kHz		
BW	Bandwidth	INA290A2Q, C _{LOAD} = 5 pF, V _{SENSE} = 80mV	700		kHz		
BW	Bandwidth	INA290A3Q, C _{LOAD} = 5 pF, V _{SENSE} = 40mV	650		kHz		
BW	Bandwidth	INA290A4Q, C _{LOAD} = 5 pF, V _{SENSE} = 20mV	600		kHz		
BW	Bandwidth	INA290A5Q, C _{LOAD} = 5 pF, V _{SENSE} = 8mV	550		kHz		
SR	Slew Rate	Rising edge, falling edge	2		V/µs		
	Settling time	V _{OUT} = 0.5 V to 4.5 V, output settles to 0.5%	9		μs		
	Settling time	V _{OUT} = 0.5 V to 4.5 V, output settles to 1%	5		μs		
NOISE	NOISE						
Ven	Voltage noise density		50		nV/\sqrt{Hz}		
POWER	POWER SUPPLY						
Vs	Supply voltage	$T_A = -40^{\circ}C \text{ to } +125^{\circ}C$	2.7	20	V		
la	Quiescent current		350	500	μΑ		
IQ	Quioscont current	$T_A = -40$ °C to +125°C		600	μΑ		

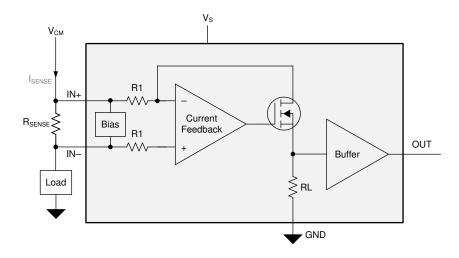


7 Detailed Description

7.1 Overview

The INA290-Q1 is a high-side only current-sense amplifier that offers a wide common-mode range, precision zero-drift topology, excellent common-mode rejection ratio (CMRR), and high bandwidth and fast slew rate. Different gain versions are available to optimize the output dynamic range based on the application. The INA290-Q1 is designed using a transconductance architecture with a current-feedback amplifier that enables low bias currents of $20~\mu\text{A}$ with a common-mode voltage of 110~V.

7.2 Functional Block Diagram





7.3 Feature Description

7.3.1 Amplifier Input Common-Mode Signal

The INA290-Q1 supports large input common-mode voltages from V_S to 110 V. Because of the internal topology, the common-mode range is not restricted by the power-supply voltage (V_S). The INA290-Q1 must be used in high-side applications where $V_{CM} > V_S$.

7.3.1.1 Low Input Bias Current

The INA290-Q1 input bias current draws 20 µA (typical) at a common-mode voltage of 110 V. This enables precision current sensing on applications that require lower leakage for high-side applications.

7.3.1.2 Low V_{SFNSF} Operation

The INA290-Q1 operates with high performance across the entire valid V_{SENSE} range. The zero-drift input architecture of the INA290-Q1 provides the low offset voltage and low offset drift needed to measure low V_{SENSE} levels accurately across the wide operating temperature of -40°C to $+125^{\circ}\text{C}$. Low V_{SENSE} operation is particularly beneficial when using low ohmic shunts for low current measurements, as power losses across the shunt are significantly reduced.

7.3.1.3 Wide Fixed Gain Output

The INA290-Q1 gain error is < 0.25% at room temperature, with a maximum drift of 10 ppm/°C over the full temperature range of - 40°C to +125°C. The INA290-Q1 is available in multiple gain options of 20 V/V, 50 V/V, 100 V/V, 200 V/V, and 500 V/V, which the system designer should select based on their desired signal-to-noise ratio and other system requirements. For example, noise-sensitive applications which require accurate measurement of low-level signals often select lower gains such as 20 V/V. Applications with high-level signals usually select higher gains such as 500 V/V.

The INA290-Q1 closed-loop gain is set by a precision, low drift internal resistor network. The ratio of these resistors are excellently matched, while the absolute values may vary significantly. Because of this variation, adding additional resistance around the INA290-Q1 to change the effective gain is not recommended. The typical values of the gain resistors are described in 表 1.

g Mod Call Robioto.				
GAIN	R1	RL		
20 (V/V)	25 kΩ	500 kΩ		
50 (V/V)	10 kΩ	500 kΩ		
100 (V/V)	10 kΩ	1000 kΩ		
200 (V/V)	5 kΩ	1000 kΩ		
500 (V/V)	2 kΩ	1000 kΩ		

表 1. Fixed Gain Resistor

7.3.1.4 Wide Supply Range

The INA290-Q1 operates with a wide supply range from a 2.7 V to 20 V. The output stage supports a full-scale output voltage range of up to V_S . Wide output range can enable very-wide dynamic range current measurements. For a gain of 20 V/V, the maximum differential input acceptable is 1 V. The offset of the INA290-Q1 is $\pm 25 \,\mu\text{V}$, and is capable of measuring a wide dynamic range of current up to 92 dB.



7.4 Device Functional Modes

7.4.1 Unidirectional Operation

The INA290-Q1 measures the differential voltage developed by current flowing through a resistor, commonly referred to as a current-sensing resistor or a current-shunt resistor. The INA290-Q1 operates in unidirectional mode only, meaning it only senses current sourced from a power supply to a system load as shown in 🛭 1.

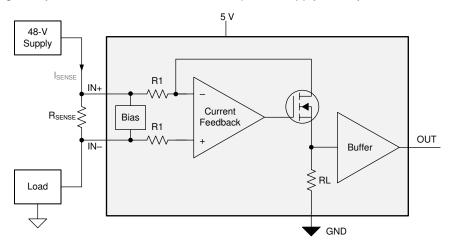


图 1. Unidirectional Application

The linear range of the output stage is limited to how close the output voltage can approach ground under zero-input conditions. The zero current output voltage of the INA290-Q1 is very small, with a maximum of GND + 25 mV. Make sure to apply a differential input voltage of (25 mV / Gain) or greater to keep the INA290-Q1 output in the linear region of operation.

7.4.2 High Signal Throughput

With a bandwidth of 800 kHz at a gain of 20 V/V and a slew rate of 2 V/ μ s, the INA290-Q1 is specifically designed for detecting and protecting applications from fast inrush currents. As shown in 表 2, the INA290-Q1 responds in less than 1 μ s for a system measuring a 75-A threshold on a 2-m Ω shunt.

表 2. Response Time

PARAMETER		EQUATION	INA290-Q1 AT V _S = 5 V
G	Gain		20 V/V
I _{MAX}	Maximum current		100 A
I _{Threshold}	Threshold current		75 A
R _{SENSE}	Current sense resistor value		2 mΩ
V _{OUT}	Output voltage	$V_{OUT} = I_{MAX} \times R_{SENSE} \times G$	4 V
SR	Slew rate		2 V/µs
	Output response time		< 1 µs



8 Application and Implementation

注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The INA290-Q1 amplifies the voltage developed across a current-sensing resistor as current flows through the resistor to the load. The wide input common-mode voltage range and high common-mode rejection of the INA290-Q1 make it usable over a wide range of voltage rails while still maintaining an accurate current measurement.

8.1.1 R_{SENSE} and Device Gain Selection

The accuracy of any current-sense amplifier is maximized by choosing the current-sense resistor to be as large as possible. A large sense resistor maximizes the differential input signal for a given amount of current flow and reduces the error contribution of the offset voltage. However, there are practical limits as to how large the current-sense resistor can be in a given application because of the resistor size and maximum allowable power dissipation. 公式 1 gives the maximum value for the current-sense resistor for a given power dissipation budget:

$$R_{SENSE} < \frac{PD_{MAX}}{{I_{MAX}}^2}$$

where:

- PD_{MAX} is the maximum allowable power dissipation in R_{SENSE}.
- I_{MAX} is the maximum current that will flow through R_{SENSE}.

An additional limitation on the size of the current-sense resistor and device gain is due to the power-supply voltage, V_S , and device swing-to-rail limitations. In order to make sure that the current-sense signal is properly passed to the output, both positive and negative output swing limitations must be examined. $\triangle \vec{x}$ 2 provides the maximum values of R_{SENSE} and GAIN to keep the device from exceeding the positive swing limitation.

where:

- I_{MAX} is the maximum current that will flow through R_{SENSE}.
- GAIN is the gain of the current-sense amplifier.
- V_{SP} is the positive output swing as specified in the data sheet.

To avoid positive output swing limitations when selecting the value of R_{SENSE}, there is always a trade-off between the value of the sense resistor and the gain of the device under consideration. If the sense resistor selected for the maximum power dissipation is too large, then it is possible to select a lower-gain device in order to avoid positive swing limitations.

The negative swing limitation places a limit on how small the sense resistor value can be for a given application. 公式 3 provides the limit on the minimum value of the sense resistor.

where:

- I_{MIN} is the minimum current that will flow through R_{SENSE}.
- · GAIN is the gain of the current-sense amplifier.
- V_{SN} is the negative output swing of the device.

(3)

表 3 shows an example of the different results obtained from using five different gain versions of the INA290-Q1. From the table data, the highest gain device allows a smaller current-shunt resistor and decreased power dissipation in the element.



Application Information (接下页)

表 3. R_{SENSE} Selection and Power Dissipation⁽¹⁾

PARAMETER		FOUNTION	RESULTS AT V _S = 5 V					
	PARAMETER	EQUATION	INA290A1Q	INA290A2Q	INA290A3Q	INA290A4Q	INA290A5Q	
G	Gain		20 V/V	50 V/V	100 V/V	200 V/V	500 V/V	
V _{SENSE}	Ideal differential input voltage	V _{SENSE} = V _{OUT} / G	250 mV	100 mV	50 mV	25 mV	10 mV	
R _{SENSE}	Current sense resistor value	R _{SENSE} = V _{SENSE} / I _{MAX}	25 mΩ	10 mΩ	5 mΩ	2.5 mΩ	1 mΩ	
P _{SENSE}	Current-sense resistor power dissipation	R _{SENSE} x I _{MAX} 2	2.5 W	1 W	0.5W	0.25 W	0.1 W	

⁽¹⁾ Design example with 10-A full-scale current with maximum output voltage set to 5 V.

8.2 Typical Application

The INA290-Q1 is a unidirectional, current-sense amplifier capable of measuring currents through a resistive shunt with shunt common-mode voltages from V_S to 110 V. The circuit configuration for monitoring current in a high-side radio frequency (RF) power amplifier (PA) application is shown in \boxtimes 2.

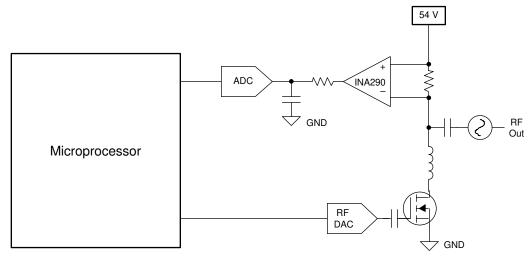


图 2. Current Sensing in a PA Application

8.2.1 Design Requirements

 V_{SUPPLY} is set to 12 V, the shunt resistor used is 50 mΩ, and the common-mode voltage set to 54 V. $\frac{1}{8}$ 4 lists the design setup for this application.

表 4. Design Parameters

DESIGN PARAMETERS	EXAMPLE VALUE
INA290 supply voltage	12 V
High-side supply voltage	54 V
Maximum sense current	5 A
R _{SENSE} resistor	50 mΩ
Gain option	50 V/V



8.2.2 Detailed Design Procedure

The INA290-Q1 measures the differential voltage across the shunt resistor, and the signal is internally amplified with a gain of 50 V/V. The output of the INA290-Q1 is connected to the analog-to-digital converter (ADC) to digitize the current measurements.

9 Power Supply Recommendations

The input circuitry of the INA290-Q1 device can accurately measure beyond the power-supply voltage. The power supply can be 20 V, whereas the load power-supply voltage at IN+ and IN- can go up to 110 V. The output voltage range of the OUT pin is limited by the voltage on the V_S pin.

10 Layout

10.1 Layout Guidelines

TI always recommends to follow good layout practices:

- Connect the input pins to the sensing resistor using a Kelvin or 4-wire connection. This connection technique
 makes sure that only the current-sensing resistor impedance is detected between the input pins. Poor routing
 of the current-sensing resistor commonly results in additional resistance present between the input pins.
 Given the very low ohmic value of the current resistor, any additional high-current carrying impedance can
 cause significant measurement errors.
- Place the power-supply bypass capacitor as close to the device power supply and ground pins as possible.
 The recommended value of this bypass capacitor is 0.1 μF. Additional decoupling capacitance can be added to compensate for noisy or high-impedance power supplies.
- When routing the connections from the current-sense resistor to the device, keep the trace lengths as short as possible.

10.2 Layout Example

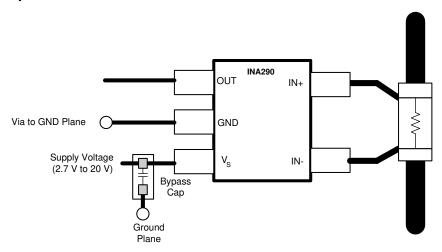


图 3. INA290-Q1 Recommended Layout



11 器件和文档支持

11.1 文档支持

11.1.1 相关文档

请参阅如下相关文档:

德州仪器 (TI), 《INA290EVM 用户指南》

11.2 接收文档更新通知

要接收文档更新通知,请导航至 ti.com. 上的器件产品文件夹。单击右上角的通知我进行注册,即可每周接收产品信息更改摘要。有关更改的详细信息,请查看任何已修订文档中包含的修订历史记录。

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11.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更,恕不另行通知,且 不会对此文档进行修订。如需获取此数据表的浏览器版本,请查阅左侧的导航栏。

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11-Mar-2021

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
INA290A1QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1G6	Samples
INA290A2QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1G8	Samples
INA290A3QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1G7	Samples
INA290A4QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1G9	Samples
INA290A5QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1GA	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.



PACKAGE OPTION ADDENDUM

11-Mar-2021

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DCK (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
- D. Falls within JEDEC MO-203 variation AA.



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