

# VARIABLE SPEED SINGLE-PHASE BLDC MOTOR CONTROLLER

### DESCRIPTION

The ZXBM1004 is a single-phase, DC brushless motor pre-driver with PWM variable speed control suitable for fan and blower motors. The controller is intended for applications where the fan or blower speed is controlled by an external PWM signal, thermistor or DC voltage.

# FEATURES

- Compliant with external PWM speed control
- Compliant with thermistor control
- Minimum speed setting
- Low noise
- Auto restart
- Built in hall amplifier
- Speed pulse (FG) and lock rotor (RD) outputs
- Up to 18V input voltage (60V with external regulator)
- QSOP16 package

### Associated application notes:-

AN41 - Thermistor control AN42 - External PWM control

AN43 - Interfacing to the motor windings

### **APPLICATIONS**

- · Mainframe and personal computer fans and blowers
- Instrumentation fans
- Central heating blowers
- Automotive climate control

### **ORDERING INFORMATION - QSOP16**

DEVICE	REEL SIZE	TAPE WIDTH	QUANTITY PER REEL
ZXBM1004Q16TA	7" (180mm)	12mm	500
ZXBM1004Q16TC	13" (330mm)	12mm	2,500

# **DEVICE MARKING**

 ZETEX ZXBM 1004

**ISSUE 6 - MAY 2007** 



**QSOP16** 



# **ABSOLUTE MAXIMUM RATINGS**

PARAMETER	SYMBOL	LIMITS	UNIT
Supply voltage	V <sub>CCmax</sub>	-0.6 to 20	V
Input current	I <sub>CCmax</sub>	200	mA
Input voltage	V <sub>IN</sub> max	-0.5 to V <sub>CC</sub> +0.5	V
Output voltage	V <sub>OUT</sub> max	-0.5 to V <sub>CC</sub> +0.5	V
Power dissipation	P <sub>Dmax</sub>	500	mW
Operating temp.	T <sub>OPR</sub>	-40 to 110	°C
Storage temp.	T <sub>STG</sub>	-55 to 150	°C

#### **Power Dissipation**

Maximum allowable Power Dissipation, P<sub>D</sub>, is shown plotted against Ambient Temperature, T<sub>A</sub>, in the accompanying Power Derating Curve, indicating the Safe Operating Area for the device.

2) Power consumed by the device, P<sub>T</sub>, can be calculated from the equation:

 $P_T = P_Q + P_{PhHi} + P_{PhLo}$ 

where  $P_Q$  is power dissipated under quiescent current conditions, given by:

 $P_Q = Vcc \times Icc$ 

- where Vcc is the application device Supply Voltage and lcc is the maximum Supply Current given in the Electrical Characteristics
- P<sub>PhHi</sub> is power generated due to either one of and the phase outputs Ph1Hi or Ph2Hi being active, given by:

 $P_{PhHi} = I_{OL} \times V_{OL}$ 

- where  $I_{OL}$  is the application Ph1Hi and Ph2Hi output currents
- $V_{0L}$  is the maximum Low Level Output Voltage for the Ph1Hi and Ph2Hi outputs given in the Electrical and Characteristics

P<sub>PhLo</sub> is power generated due to either one of the phase outputs Ph1Lo or Ph2Lo being and active, given by:

 $P_{PhLo} = I_{OH} \times (V_{CC} - V_{OH})$ 

- where IOH is the application Ph1Lo and Ph2Lo output currents
- and Vcc is the application device Supply Voltage
- V<sub>OH</sub> is the minimum High Level and Output Voltage for the Ph1Lo and Ph2Lo outputs given in the Electrical Characteristics



**Power Derating Curve** 



PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT	CONDITIONS
Supply voltage	V <sub>CC</sub>	4.7		18	V	
Supply current	I <sub>CC</sub>		6	8.5	mA	No load <sup>(1)</sup>
Hall amp input voltage	VIN	40			mV	diff p-p
Hall amp common mode voltage	V <sub>CM</sub>	0.5		V <sub>CC</sub> -1.5	V	
Hall amp input offset	V <sub>OFS</sub>		±7		mV	
Hall amp bias current	I <sub>BS</sub>		400	650	nA	
Ph1Lo, Ph2Lo output high voltage	V <sub>OH</sub>	V <sub>CC</sub> -2.2	V <sub>CC</sub> -1.8		V	I <sub>OH</sub> =80mA
Ph1Lo, Ph2Lo output low voltage	V <sub>OLA</sub>		0.4	0.6	V	I <sub>OL</sub> =16mA <sup>(2)</sup>
Ph1Lo, Ph2Lo output low voltage	V <sub>OLB</sub>		0.4	0.6	V	I <sub>OL</sub> =50μA <sup>(3)</sup>
Ph1Lo, Ph2Lo output source current	I <sub>OH</sub>			-80	mA	
Ph1Lo, Ph2Lo output sink current	I <sub>OL</sub>			16	mA	
Ph1Hi, Ph2Hi output low voltage	V <sub>OL</sub>		0.5	0.7	V	I <sub>OL</sub> =100mA
Ph1Hi, Ph2Hi output sink current	I <sub>OL</sub>			100	mA	
C <sub>PWM</sub> charge current	IPWMC	-6.2	-7.6	-9	μA	
C <sub>PWM</sub> discharge current	I <sub>PWMD</sub>	65	80	95	μA	
C <sub>PWM</sub> high threshold voltage	V <sub>THH</sub>	2.95	3	3.15	V	
C <sub>PWM</sub> low threshold voltage	V <sub>THL</sub>	0.94	1	1.11	V	
PWM frequency	F <sub>PWM</sub>		24		kHz	C <sub>PWM</sub> =0.1nF
ThRef voltage	V <sub>ThRef</sub>	2.925	3.0	3.14	V	I <sub>OThRef</sub> =-100μA
ThRef output current	IOThRef		-0.2	-1	mA	
S <sub>MIN</sub> input current	IISMIN		-0.25	-0.5	μA	Vin = 2V
SPD voltage minimum	V <sub>SPDL</sub>		1		V	100% PWM drive
SPD voltage maximum	V <sub>SPDH</sub>		3		V	0% PWM drive
SPD input current	I <sub>ISPD</sub>		-0.8	-2	μΑ	Vin = 2V
C <sub>LCK</sub> charge current	ILCKC	-2	-3		μA	
C <sub>LCK</sub> discharge current	ILCKD		0.2	0.35	μA	
C <sub>LCK</sub> high threshold voltage	V <sub>THH</sub>		3		V	
C <sub>LCK</sub> low threshold voltage	V <sub>THL</sub>		1		V	
Lock condition On:Off ratio			1:12			
FG & RD output current	I <sub>OL</sub>			5	mA	
FG & RD low level output voltage	V <sub>OL</sub>			0.5	V	I <sub>OI</sub> =5mA

# **ELECTRICAL CHARACTERISTICS** (at $T_{amb} = 25^{\circ}C \& V_{CC} = 12V$ )

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Notes: (1) Measured with pins H+, H-,  $C_{LCK}$  and  $C_{PWM} = 0V$  and all other signal pins open circuit. (2) Measured when opposing phase output is low. (3) Measured when opposing phase output is high.





# Block diagram



# Pin assignments





### PIN FUNCTIONAL DESCRIPTION

# H+ - Hall input H- - Hall input

The rotor position is detected by a Hall sensor, with the output applied to the H+ and H- pins. This sensor can be either a 4 pin 'naked' Hall device or of the 3 pin buffered switching type. For a 4 pin device the differential Hall output signal is connected to the H+ and H- pins. For a buffered Hall sensor the Hall device output is attached to the H+ pin, with a pull-up attached if needed, whilst the H- pin has an external potential divider attached to hold the pin at half Vcc. When H+ is high in relation to H-, Ph2 is the active drive.

#### **ThRef - Network Reference**

This is a reference voltage of nominal 3V. It is designed for the ability to 'source' and therefore it will not 'sink' any current from a higher voltage.

The total current drawn from the pin by the minimum speed potential divider to pin S<sub>MIN</sub> and any voltage setting network should not exceed 1mA at maximum temperature.

#### **SPD** - Speed Control Input

The voltage applied to the SPD pin provides control over the Fan Motor speed by varying the Pulse Width Modulated (PWM) drive ratio at the Ph1Lo and Ph2Lo outputs. The control signal takes the form of a voltage input of range 3V to 1V, representing 0% to 100% drive respectively.

If variable speed control is not required this pin can be left with an external potential divider to set a fixed speed or tied to ground to provide full speed i.e. 100% PWM drive.

If required this pin can also be used as an enable pin. The application of a voltage >3.0V will force the PWM drive fully off, in effect disabling the drive.

### S<sub>MIN</sub> - Sets Minimum Speed

A voltage can be set on this pin via a potential divider between the ThRef and Gnd. This voltage is monitored by the SPD pin such that it cannot rise above it. As a higher voltage on the SPD pin represents a lower speed it therefore restricts the lower speed range of the fan. If this feature is not required the pin is left tied to ThRef so no minimum speed will be set.

If the fan is being controlled from an external voltage source onto the SPD pin then either this feature should not be used or if it is required then a >1k $\Omega$  resistor should be placed in series with the SPD pin.

### C<sub>PWM</sub> - Sets PWM Frequency

This pin has an external capacitor attached to set the PWM frequency for the Phase drive outputs. A capacitor value of 0.1nF will provide a PWM frequency of typically 24kHz.

The CPWM timing period (TPWM) is determined by the following equation:

$$T_{PWM} = \frac{(V_{THH} - V_{THL}) \times C}{I_{PWMC}} + \frac{(V_{THH} - V_{THL}) \times C}{I_{PWMD}}$$

 $C = C_{PWM} + 15$ , in pF Where:  $V_{THH}$  and  $V_{THL}$  are the  $C_{PWM}$  pin threshold voltages IPWMC and IPWMD are the charge and discharge currents in µA. T<sub>PWM</sub> is in ms



As these threshold voltages are nominally set to  $V_{THH}$  = 3V and  $V_{THL}$  = 1V the equations can be simplified as follows:

$$T_{PWM} = \frac{2C}{I_{PWMC}} + \frac{2C}{I_{PWMD}}$$

#### C<sub>LCK</sub> - Locked rotor timing capacitor

Should the fan stop rotating for any reason, i.e. an obstruction in the fan blade or a seized bearing, then the device will enter a Rotor Locked condition. In this condition after a predetermined time ( $T_{LOCK}$ ) the RD pin will go high and the Phase outputs will be disabled. After a further delay ( $T_{OFF}$ ) the controller will re-enable the Phase drive for a defined period ( $T_{ON}$ ) in an attempt to re-start the fan. This cycle of ( $T_{OFF}$ ) and ( $T_{ON}$ ) will be repeated indefinitely or until the fan re-starts.

### **GND** - Ground

This is the device supply ground return pin and will generally be the most negative supply pin to the fan.

#### **RD** - Locked Rotor error output

This pin is the Locked Rotor output as referred to in the  $C_{LCK}$  timing section above. It is high when the rotor is stopped and low when it is running.

This is an open collector drive giving an active pull down with the high level being provided by an external pull up resistor.

#### FG - Frequency Generator (speed) output

This is the Frequency Generator output and is a buffered signal from the Hall sensor.

This is an open collector drive giving an active pull down with the high level being provided by an external pull up resistor.

### Ph1Lo & Ph2Lo - Low-side External H-bridge Driver

This pair of outputs drive the Low side of the external high power H-bridge devices which in turn drives the single phase winding. These outputs provide both the commutation and PWM waveforms. The outputs are of the Darlington emitter follower type with an active pull-down to help faster switch off when using bipolar devices. When in the high state the outputs will provide up to 80mA of drive into the base or gates of external transistors as shown in the Typical Application circuit following.

When in the low state the active Phase drive is capable of sinking up to 16mA when driving low to aid turn off times during PWM operation. When the Phase is inactive the output is held low by an internal pull-down resistor.

#### Ph1Hi & Ph2Hi - High-side External H-bridge Driver

These are the High side outputs to the external H-bridge and are open collector outputs capable of sinking 100mA. This signal provides commutation only to the H-bridge.

### V+OP - Phase Outputs Supply Voltage

This pin is the supply to the Phase outputs and will be connected differently dependant upon external transistor type.

For bipolar devices this pin will be connected by a resistor to the  $V_{CC}$  pin. The resistor is used to control the current into the transistor base so its value is chosen accordingly.

For MOSFET devices the pin will connect directly to the  $V_{\text{CC}}$  pin.



### V<sub>CC</sub> - Applied Voltage

This is the device internal circuitry supply voltage. For 5V to 12V fans this can be supplied directly from the Fan Motor supply. For fans likely to run in excess of the 18V maximum rating for the device this will be supplied from an external regulator such as a Zener diode.

### **RD Timing Waveform:**



### **Applications Information**

The ZXBM1004 is primarily controlled by a voltage on the SPD pin. A voltage of 1V represents a 100% PWM at the Phase Outputs and in turn represents full speed. 3V on the SPD pin conversely represents 0% PWM. The motor can therefore be controlled simply by applying a control voltage onto the SPD pin with the minimal use of external components.

This voltage control method easily lends itself to control by other signal types. For example if a thermistor is applied to the SPD pin a varying voltage can be generated at the SPD pin as the resistance of the thermistor varies with temperature. A common form of control of fans is by a PWM signal derived from a central processor or controller. This signal can be converted into a voltage and that voltage adjusted as neccesary to compensate for motor none linearity, inclusion of the Minimum speed feature etc.

Full applications details and further examples of how to control the ZXBM1004 are available in the Applications Notes AN41, AN42 and AN43.





Figure 1: Typical circuit for thermistor controlled speed



Figure 2: Typical circuit for external PWM controlled speed





Figure 3: Typical circuit for 48V input and external PWM control



Figure 4: Typical circuit for constant speed operation



### **Drive transisitors**

Zetex offers a range of devices that are ideally suited to interface between the ZXBM1004 controller and the motor. The following tables show a selection of products ranging from single packaged H-bridge devices to individual power components that can be used in this application. If your needs are not covered by this selection then please refer to the the more comprehensive listings that can be found on the Zetex website: www.zetex.com

Description	Part Number	Content	BV <sub>DSS</sub>	I <sub>D</sub>	I <sub>DM</sub>	R <sub>DS(on)</sub> @V <sub>GS</sub> =10V	Package
			V	Α	Α	max. m $\Omega$	
MOSFET	ZXMHC6A07T8	2 x N	60	1.8	8.7	300	SM-8
H-Bridge		2 x P	-60	-1.5	-7.5	425	
	ZXMHC3A01T8	2 x N	30	3.1	14.5	120	SM-8
		2 x P	-30	-2.3	-10.8	210	
N + P	ZXMC4559DN8	N	60	4.7	22	55	SO8
channel MOSFET		Р	-60	-3.5	-18	105	
	ZXMC3A16DN8	N	30	6.4	30	35	SO8
		Р	-30	-5.4	-25	48	
Low side	ZXMN10A09K	N	100	7.7	27	85	
switch MOSFET	ZXMN6A09K	N	60	11.2	40	45	DPAK
	ZXMN3A04K	N	30	18.4	66	20	

Description	Part Number	Content	BV <sub>CEO</sub> V	l <sub>c</sub> A	I <sub>см</sub> А	V <sub>CE(sat)</sub> at I <sub>C</sub> =2A max. mV	Package
High side	ZXT953K	PNP	-100	-5	-10	-175	
switch Bipolar	ZXT951K	PNP	-60	-6	-15	-165	DPAK
ыротаг	ZXT790AK	PNP	-40	-3	-6	-450	



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#### Product status key:

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"Last time buy (LTB)"Device will be discontinued and last time buy period and delivery is in effect

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## PACKAGE OUTLINE QSOP16



Dim	Inche	s	Millir	neters	
	Min.	Max.	Min.	Max.	
А	0.053	0.069	1.35	1.75	
A1	0.004	0.010	0.10	0.25	
A2	0.049	0.059	1.25	1.50	
D	0.189	0.197	4.80	5.00	
ZD	0.009	0.009 Ref		BSC	
E	0.228	0.244	5.79	6.20	
E1	0.150	0.157	3.81	3.99	
L	0.016	0.050	0.41	1.27	
е	0.025 E	SC	0.64	BSC	
b	0.008	0.012	0.20	0.30	
С	0.007	0.010	0.18	0.25	
θ	0°	8°	0°	8°	

# PACKAGE DIMENSIONS

Note: Dimensions in inches are control dimensions, dimensions in millimeters are approximate.

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